

# Complex Impedance

We have previously represented sinusoidal signals in the form of:

$$V_a(t) = A \sin(\omega t) \quad \text{where } B, A = \text{amplitude}$$

$$V_{out}(t) = B \sin(\omega t + \phi) \quad \begin{array}{l} \omega = \text{frequency} \\ \phi = \text{phase} \end{array}$$

Now, we will represent voltages using the form:

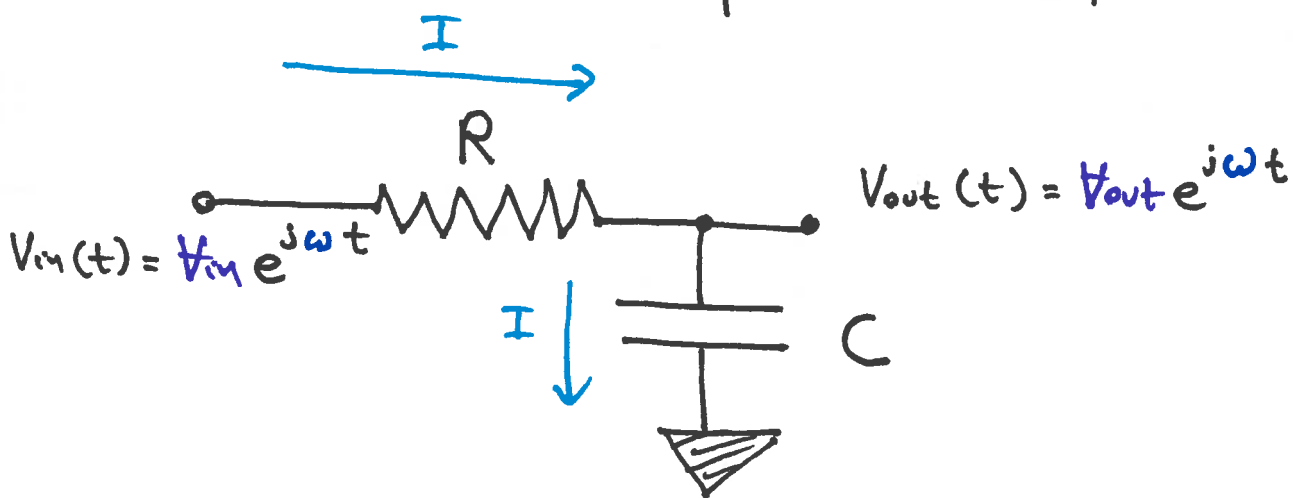
$$V(t) = \mathcal{V} e^{j\omega t}$$

where  $\mathcal{V}$  is a complex amplitude of the form:

$$\mathcal{V} = x + jy \quad \text{or} \quad \mathcal{V} = r e^{j\phi}$$

$$\text{so, } v(t) = r e^{j(\omega t + \phi)}$$

Consider the low pass filter:



$$V_{in}(t) - V_{out}(t) = IR$$

$$I = C \frac{d(V_{out}(t) - 0)}{dt}$$

Setting the currents equal:

$$\frac{V_{in}(t) - V_{out}(t)}{R} = C \frac{dV_{out}(t)}{dt}$$

$$\frac{V_{in} e^{j\omega t} - V_{out} e^{j\omega t}}{R} = C j\omega V_{out} e^{j\omega t}$$

$$\frac{(V_{in} - V_{out}) e^{j\omega t}}{R} = C j\omega V_{out} e^{j\omega t}$$

$$\frac{V_{in} - V_{out}}{R} = C j\omega V_{out}$$

$$\boxed{\frac{V_{out}}{V_{in}} = \frac{1}{j\omega RC + 1}}$$

Now how about magnitude and phase?

$$\frac{V_{out}}{V_{in}} = \frac{1}{1 + j\omega RC} \cdot \frac{1 - j\omega RC}{1 - j\omega RC} = \frac{1 - j\omega RC}{1 + \omega^2 R^2 C^2}$$

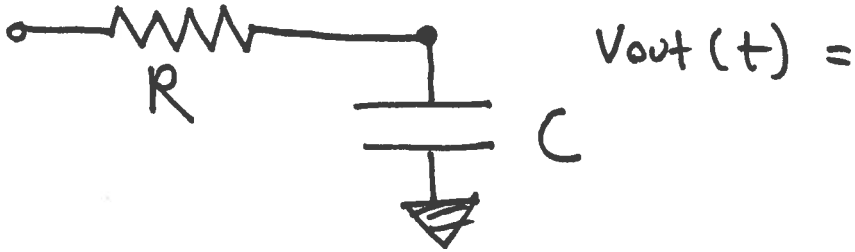
Recall we can write complex numbers in the form:

$$\frac{V_{out}}{V_{in}} = x + jy = r e^{j\phi}$$

$$\text{where } r = \sqrt{x^2 + y^2} \quad \text{and } \phi = \tan^{-1}\left(\frac{y}{x}\right)$$

Consider the low pass filter:

$$V_{in}(t) =$$



In the polar form:

$$x = \frac{1}{1 + \omega^2 R^2 C^2}$$

$$y = \frac{-RC\omega}{1 + \omega^2 R^2 C^2}$$

Then, we have:

$$r = \left[ \frac{1}{(1 + \omega^2 R^2 C^2)^2} + \frac{R^2 C^2 \omega^2}{(1 + \omega^2 R^2 C^2)^2} \right]^{1/2} = \frac{1}{\sqrt{1 + \omega^2 R^2 C^2}}$$

$$r = \frac{1}{\sqrt{1 + (\omega RC)^2}}$$

$$\phi = \tan^{-1} \left( \frac{\left( \frac{-RC\omega}{1 + \omega^2 R^2 C^2} \right)}{\left( \frac{1}{1 + \omega^2 R^2 C^2} \right)} \right) = \tan^{-1}(-RC\omega)$$

$$\phi = -\tan^{-1}(\omega RC)$$

So, for the low pass filter:

$$\frac{V_{out}}{V_{in}} = \frac{1}{1 + j\omega RC} = \frac{1}{\sqrt{1 + (\omega RC)^2}} e^{-j \tan^{-1}(\omega RC)}$$

This is the same result found using  
Sines and cosines in Chapter 4

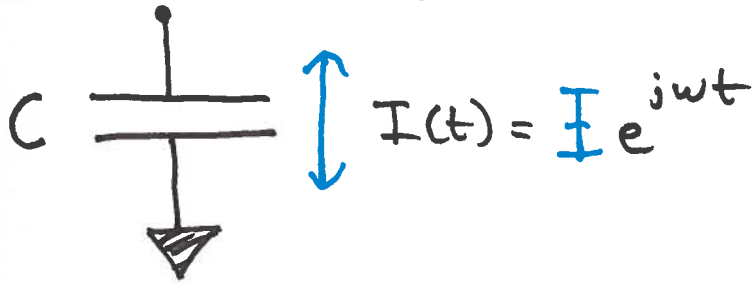
Solve for  $r$  and  $\phi$  for:

$$\frac{V_{out}}{V_{in}} = \frac{1}{1 + j\omega RC}$$

Ok, can we generalize this result?

consider the capacitor:

$$V(t) = V_m e^{j\omega t}$$



$\Rightarrow$  complex representation of current

$$I(t) = C \frac{dV(t)}{dt} = \frac{C d V_m e^{j\omega t}}{dt}$$

$$I e^{j\omega t} = C j\omega V_m e^{j\omega t}$$

$$\boxed{\frac{V_m}{I} = \frac{1}{j\omega C}} \Rightarrow \text{similar to } \frac{V}{I} = R !!!$$

So, we can now consider the concept of an Impedance being the complex equivalent of a resistance

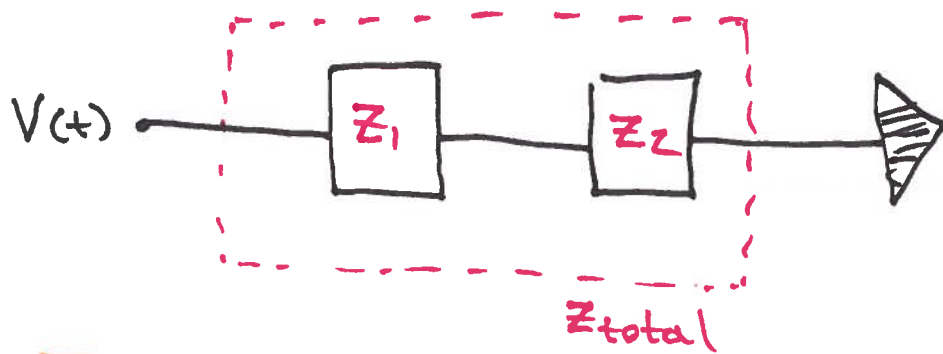
For a capacitor

$$Z_C = \frac{1}{j\omega C}$$

For a resistor

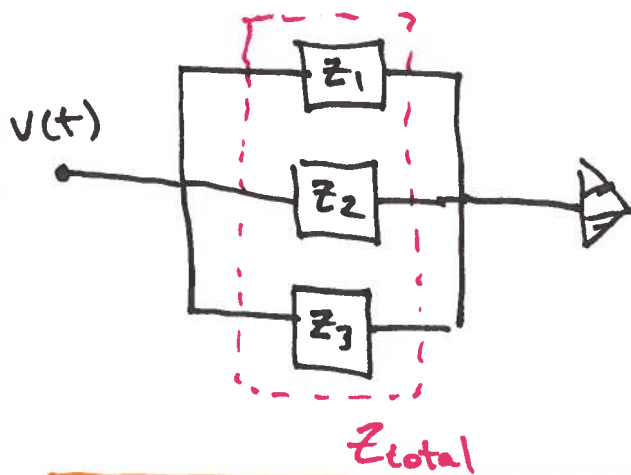
$$Z_R = R$$

Impedance blocks act like resistors!



$$Z_{total} = Z_1 + Z_2 + \dots + Z_n$$

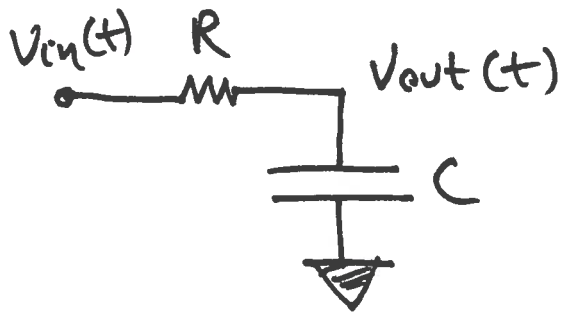
Impedances in series add!



$$Z_{total} = \frac{1}{\frac{1}{Z_1} + \frac{1}{Z_2} + \frac{1}{Z_3} + \dots + \frac{1}{Z_n}}$$

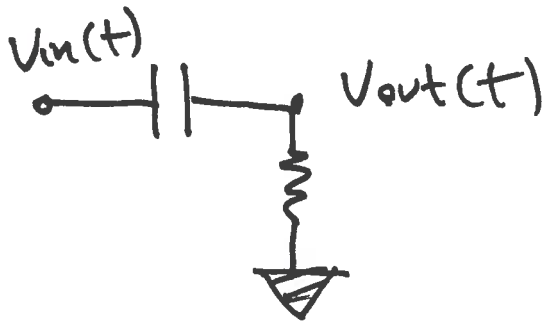
Impedances in parallel act like resistors in parallel!

Use **Impedances** to solve the low pass filter:

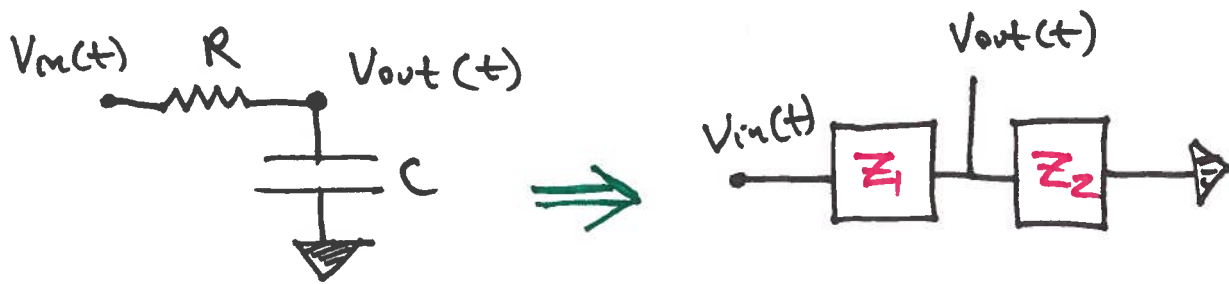




Use Impedances to solve the high pass filter



Let's apply the concept of **Impedance** to our high and low pass filters



$$\frac{V_{in}(t) - V_{out}(t)}{Z_1} = \frac{V_{out}(t) - 0}{Z_2}$$

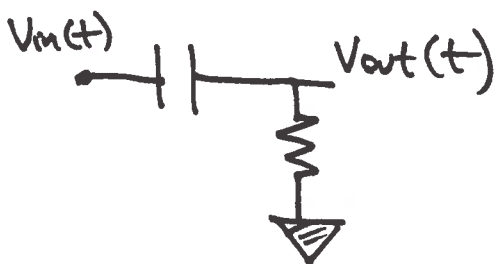
$$\Rightarrow \frac{V_{out} e^{j\omega t}}{V_{in} e^{j\omega t}} = \frac{Z_2}{Z_1 + Z_2}$$

$$\Rightarrow \frac{V_{out}}{V_{in}} = \frac{Z_2}{Z_1 + Z_2} \quad \text{This is a voltage divider!}$$

For a low pass filter:  $Z_1 = R$ ,  $Z_2 = \frac{1}{j\omega C}$

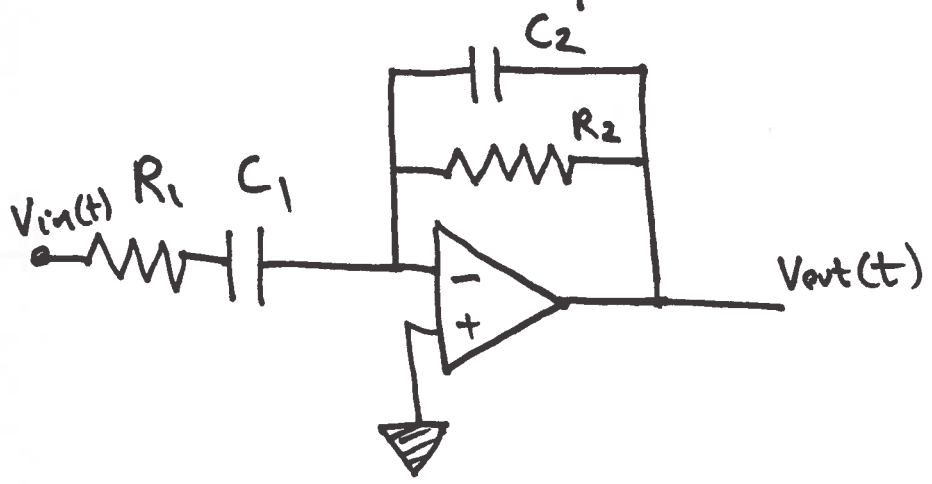
$$\frac{V_{out}}{V_{in}} = \frac{1/j\omega C}{R + 1/j\omega C} = \frac{1}{1 + j\omega RC}$$

For a high pass filter:  $Z_1 = \frac{1}{j\omega C}$ ,  $Z_2 = R$

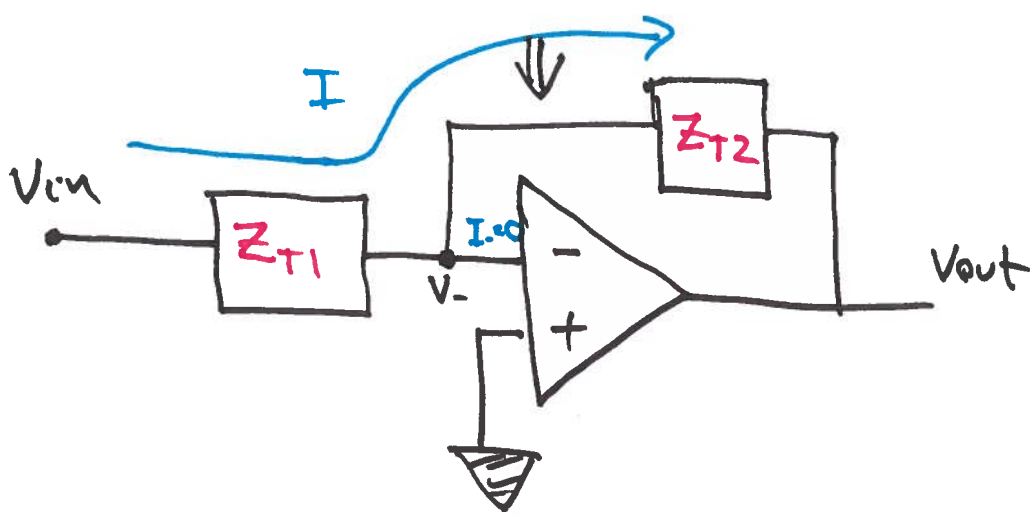
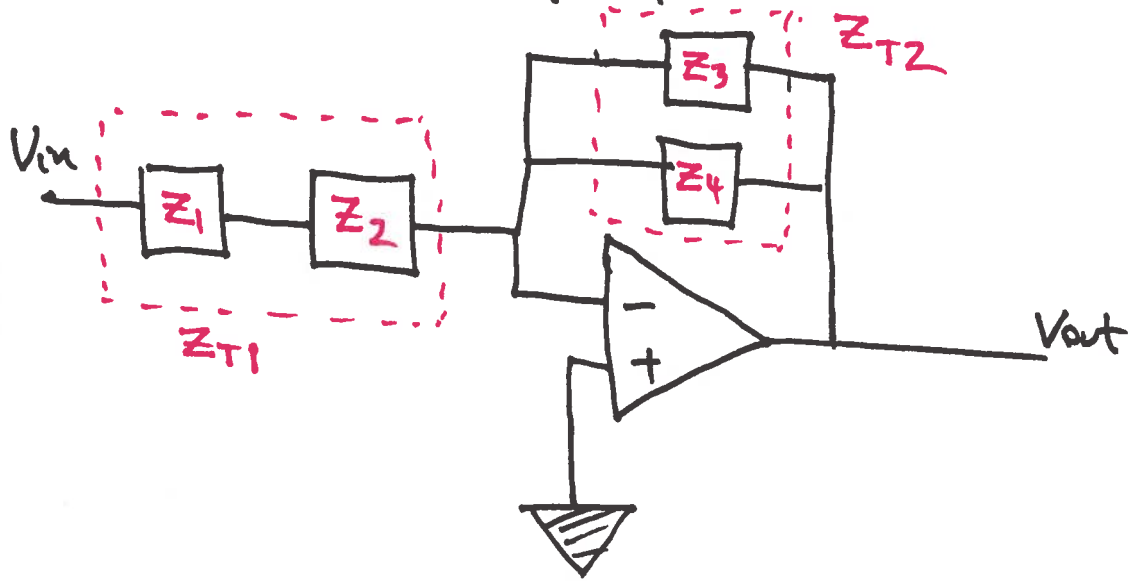


$$\frac{V_{out}}{V_{in}} = \frac{Z_2}{Z_1 + Z_2} = \frac{j\omega RC}{1 + j\omega RC}$$

Consider the bandpass filter:



↓ represent with impedances



$$Z_{T1} = Z_1 + Z_2 = R_1 + \frac{1}{j\omega C_1}$$

$$Z_{T2} = \frac{1}{1/Z_3 + 1/Z_4}$$

$$Z_{T1} = Z_1 + Z_2 = R_1 + \frac{1}{j\omega C_1}$$

$$Z_{T1} = R_1 + \frac{1}{j\omega C_1}$$

$$Z_{T2} = \frac{1}{\frac{1}{Z_3} + \frac{1}{Z_4}} = \frac{R_2}{1 + j\omega C_2 R_2}$$

$$Z_{T2} = \frac{R_2}{1 + j\omega C_2 R_2}$$

Now, Apply the op-amp rules:  $I_- = 0$

$$V_{in}(t) - V_- = I Z_{T1} \quad \text{and} \quad V_- - V_{out}(t) = I Z_{T2}$$

With the reference to ground,  $V_- = 0$

$$\frac{V_{in}(t) - 0}{Z_{T1}} = \frac{0 - V_{out}(t)}{Z_{T2}}$$

$$\frac{V_{out}(t)}{V_{in}(t)} = \frac{V_{out}}{V_{in}} = -\frac{Z_{T2}}{Z_{T1}}$$

The analysis using impedance blocks is super quick!

We can now substitute what we found for  $Z_{T1}$  and  $Z_{T2}$  and learn more about this circuit

$$\frac{V_{out}}{V_{in}} = \underbrace{\frac{-R_2}{1+j\omega C_2 R_2}}_{Z_{T1}} \cdot \underbrace{\frac{1}{R_1 + \frac{1}{j\omega C_1}}}_{Z_{T2}}$$

$$= \frac{-R_2}{1+j\omega C_2 R_2} \cdot \frac{j\omega C_1}{1+j\omega C_1 R_1}$$

$$\frac{V_{out}}{V_{in}} = \frac{-R_2}{R_1} \cdot \frac{1}{1+j\omega C_2 R_2} \cdot \frac{j\omega R_1 C_1}{1+j\omega R_1 C_1}$$

What does this result tell us about the circuit

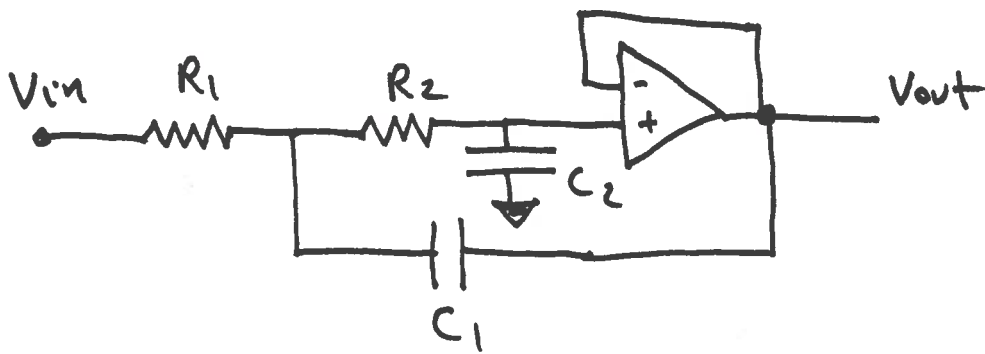
$$\frac{V_{out}}{V_{in}} = \frac{-R_2}{R_1} \cdot \frac{1}{1+j\omega C_2 R_2} \cdot \frac{j\omega R_1 C_1}{1+j\omega R_1 C_1}$$

↗  
Amplifier with gain  $\frac{R_2}{R_1}$

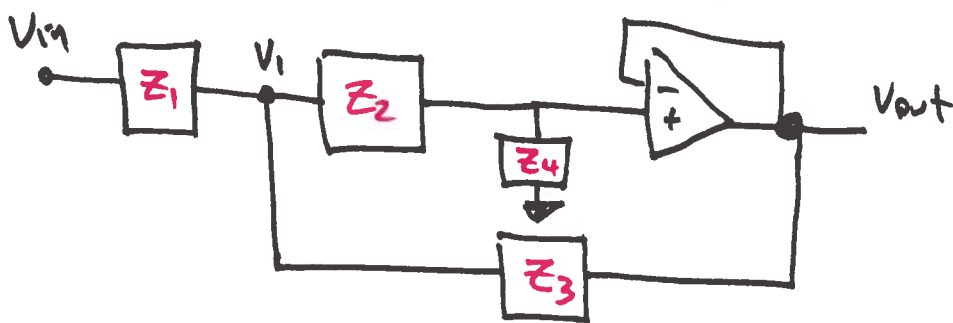
↑↑  
Low pass filter  
 $\omega_c = \frac{1}{R_2 C_2}$

↑↑  
High pass filter  
 $\omega_c = \frac{1}{R_1 C_1}$

# Sallen-key Topology



⇓



$$\underline{V_+ = V_- = V_{out}}$$

Applying op-amp rules:

$$\frac{V_{in} - V_1}{Z_1} = \frac{V_1 - V_{out}}{Z_3} + \frac{V_1 - V_{out}}{Z_4}$$

and

$$\frac{V_1 - V_{out}}{Z_2} = \frac{V_{out}}{Z_4}$$

$$\frac{V_{out}}{V_{in}} = \frac{Z_3 Z_4}{Z_1 Z_2 + Z_3(Z_1 + Z_2) + Z_3 Z_4} \quad \begin{matrix} \text{let } Z_1 = Z_2 \\ Z_3 = Z_4 \end{matrix} = \frac{Z_3^2}{(Z_1 + Z_3)^2}$$

$$\frac{V_{out}}{V_{in}} = \frac{(1/j\omega C)^2}{(R + 1/j\omega C)^2} \Rightarrow \text{2nd order low pass}$$